

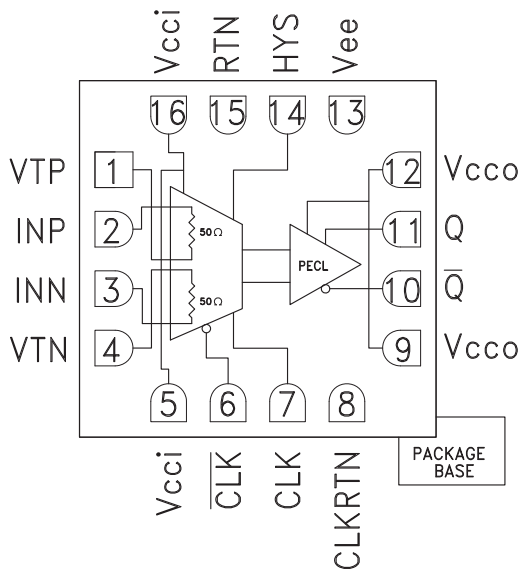
## 20 Gbps Clocked COMPARATOR with RSCML OUTPUT STAGE

### Typical Applications

The HMC875LC3C is ideal for:

- ATE Applications
- High Speed Instrumentation
- Digital Receiver Systems
- Pulse Spectroscopy
- High Speed Trigger Circuits
- Clock & Data Restoration

### Functional Diagram



### Features

- Propagation Delay Clock to Output: 120 ps
- Overdrive & Slew Rate Dispersion: 10 ps
- Minimum Pulse Width: 60 ps
- Resistor Programmable Hysteresis
- Differential Clock Control
- Input Bandwidth: 10 GHz
- Power Dissipation: 130 mW
- RSPECL and RSECL Versions Available
- 16 Lead 3x3mm SMT Package: 9mm<sup>2</sup>

### General Description

The HMC875LC3C is a SiGe monolithic, ultra fast comparator which features reduced swing CML output drivers and clock inputs. The comparator supports 20 Gbps operation while providing 120 ps clock to data output delay and 60 ps minimum pulse width with 0.2 ps rms random jitter (RJ). 25 Gbps operation can be achieved with reduced output voltage swing. Overdrive and slew rate dispersion are typically 10 ps, making the device ideal for a wide range of applications from ATE to broadband communications. The reduced swing CML output stage is designed to directly drive 400 mV into 50 Ohms terminated to GND. The HMC875LC3C features high speed latches with programmable hysteresis, and is configured to operate as a clocked comparator.

### Electrical Specifications

$T_A = +25^\circ\text{C}$   $V_{CCI} = +3.3\text{V}$ ,  $V_{CCO} = 0\text{V}$ ,  $\overline{\text{CLK}} / \text{CLK} = 1.6\text{V to } 2.4\text{V}$ ,  $V_{EE} = -3\text{V}$ ,  $V_{TERM} = 0\text{V}$

Parameter	Conditions	Min.	Typ.	Max	Units
Input Voltage Range		-2		2	V
Input Differential Voltage		-1.75		1.75	V
Input Offset Voltage			±5		mV
Input Offset Voltage, Temperature Coefficient			15		μV / °C
Input Bias Current			15		μA
Input Bias Current Temperature Coefficient			50		nA / °C
Input Offset Current			4		μA
Input Impedance			50		Ω
Common Mode Input Impedance			350		KΩ
Differential Input Impedance			15		KΩ
Hysteresis	$R_{hys} = \infty$		±1		mV


**20 Gbps Clocked COMPARATOR  
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**Latch Enable Characteristics**

Parameter	Conditions	Min.	Typ.	Max	Units
Clock Enable Input Impedance	Each Pin		50		$\Omega$
Clock to Data Output Delay, tpd			120		ps
Clock Enable Input Range		1.6		2.4	V
Clock Max Frequency, fmax			25		GHz

**DC Output Characteristics,  $V_{CCO} = 0V$  with  $50\Omega$  to  $V_{TERM} = 0V$** 

Parameter	Conditions	Min.	Typ.	Max	Units
Output Voltage High Level, Voh		-10		0	mV
Output Voltage Low Level, Vol		-420		-330	mV
Output Voltage Differential Swing		330		420	mV

**AC Performance**

Parameter	Conditions	Min.	Typ.	Max	Units
VOD Dispersion	$50mV < VOD < 1V$		10		ps
Tpd vs. Common Mode Dispersion, $-1.75V < V_{cm} < 1.75V$	$VOD = 50mV$		3		ps
Equivalent Input Bandwidth <sup>[1]</sup>		8.3	9.6	11.2	GHz
Deterministic Jitter (pp)	Deterministic Jitter at 10 Gbps with $\pm 100mV$ Overdrive		< 3		ps
Random Jitter (rms)	Random Jitter at 10 Gbps with $\pm 100mV$ Overdrive		0.2		ps rms
Minimum Pulse Width			60		ps
Q / $\bar{Q}$ Rise Time	From 20% to 80%		28		ps
Q / $\bar{Q}$ Fall Time	From 20% to 80%		22		ps

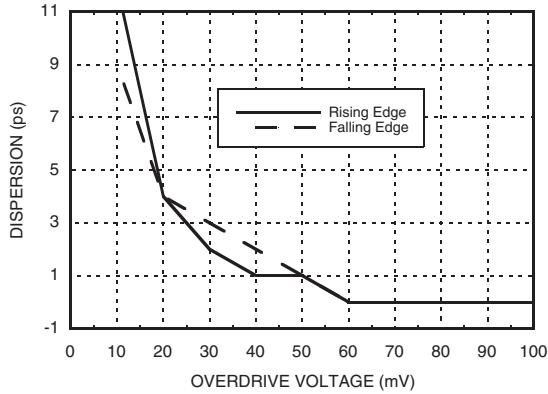
**Power Supply Requirements**

Parameter	Conditions	Min.	Typ.	Max	Units
Input Supply Current, I <sub>cci</sub>			12		mA
Output Supply Current, I <sub>cco</sub>			9		mA
Vee Current, I <sub>ee</sub>			29		mA
Power Dissipation, Pd			130		mW
PSRR, V <sub>cci</sub>			35		dB
PSRR, V <sub>ee</sub>			35		dB

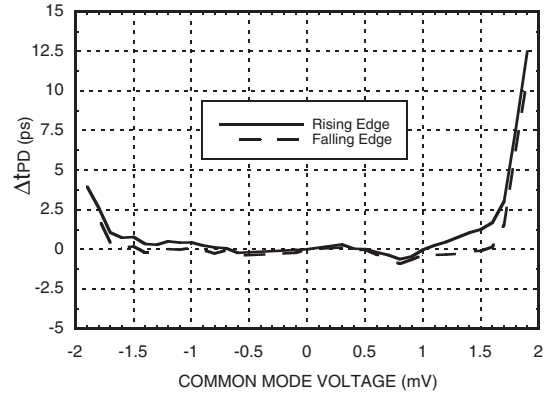
Note 1: Equivalent Input Bandwidth is calculated with the following formula:  $B_{weq} = 0.22 / J$  (TRCOMP<sup>2</sup> - TRIN<sup>2</sup>) where TRIN is the 20%/80% transition time of a quasi-Gaussian signal applied to the comparator input, and TRCOMP is the effective transition time digitized by the comparator.

## 20 Gbps Clocked COMPARATOR with RSCML OUTPUT STAGE

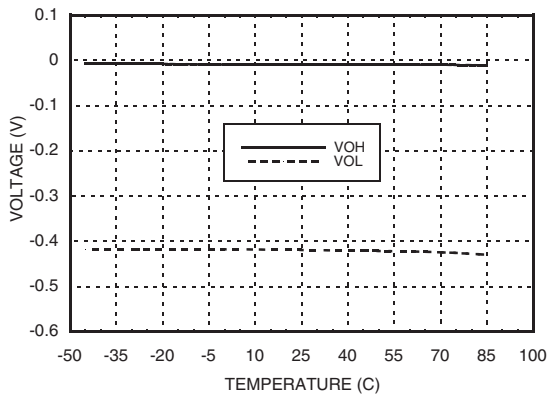
**Dispersion vs. Overdrive Voltage**



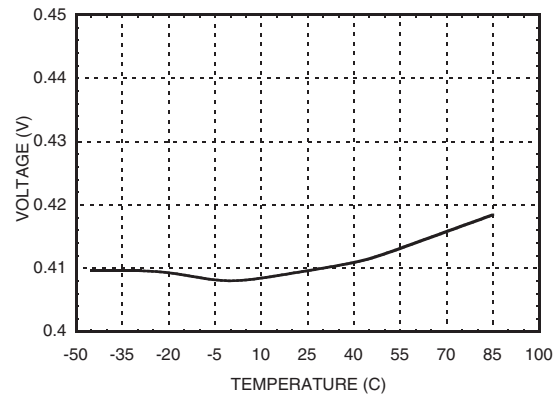
**Propagation Delay vs. Input Common Mode Voltage<sup>[1]</sup>**



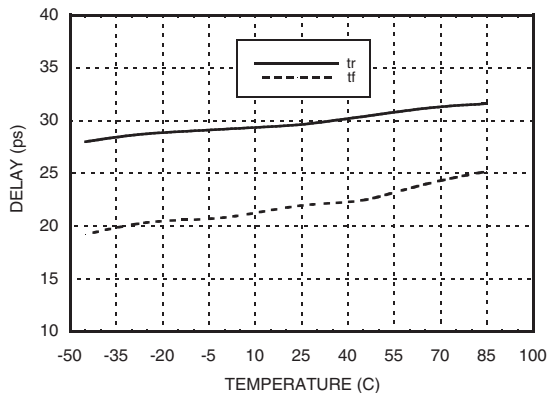
**Output Voltage vs. Temperature**



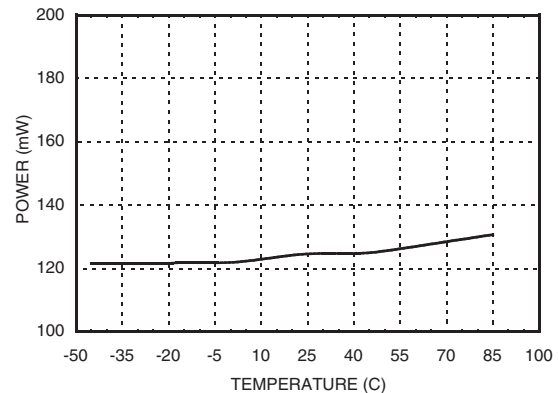
**Voltage Swing vs. Temperature**



**Delay vs. Temperature**



**Power Dissipation vs. Temperature**

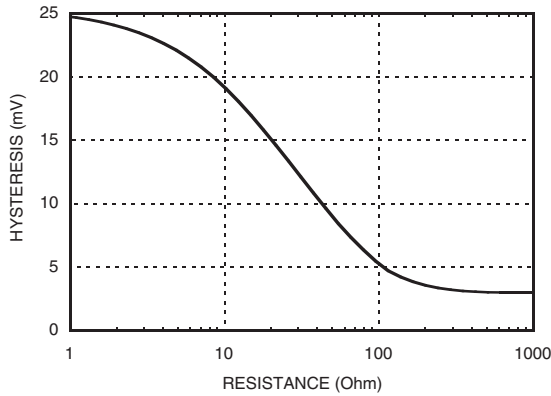


[1]  $V_{CCI} = +3.3V$ ,  $V_{CCO} = 0V$ ,  $V_{EE} = -3V$ ,  $V_{TERM} = 0V$

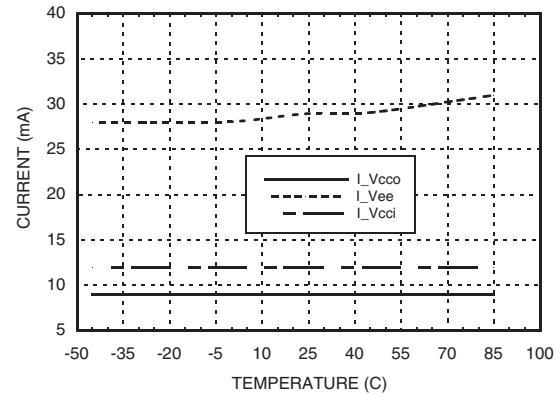


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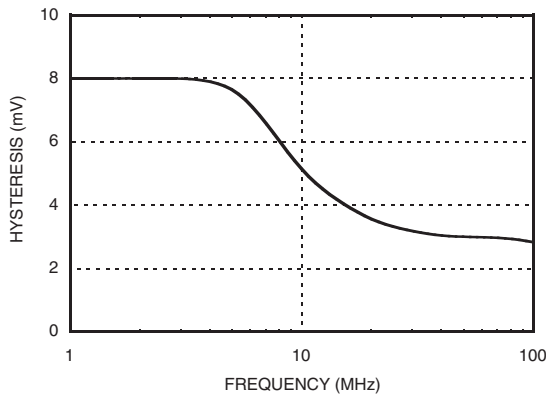
### Comparator Hysteresis vs. Rhys Control Resistor



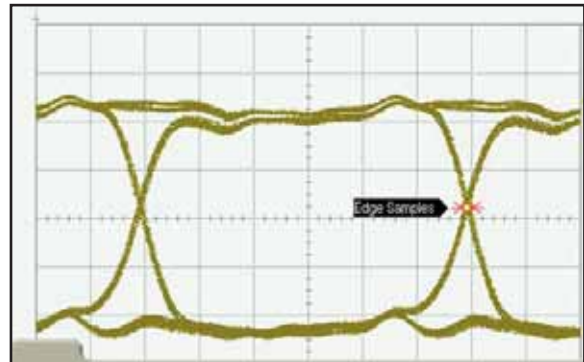
### Currents vs. Temperature



### Comparator Hysteresis vs. Clock Frequency (Rhys = ∞)



### Eye Diagram



TJ(1E-12):	6.71 ps	DJ(δ-δ):	3.08 ps	RJ(rms):	265 fs
RJ(δ-δ):	310 fs	DDJ(p-p):	3.24 ps	DCD:	-----
PJ(rms)	0.0 s			ISI J(p-p)	3.24 ps

### Absolute Maximum Ratings

Input Supply Voltage (Vcci to GND)	-0.5V to +4V
Output Supply Voltage (Vcco to GND)	-0.5V to +4V
Positive Supply Differential (Vcci - Vcco)	-0.5V to +3.5V
Input Voltage	-2V to +2V
Differential Input Voltage	-2V to +2V
Input Voltage, Latch Enable	-0.5V to Vcci +0.5V
Applied Voltage (HYS)	Vee to GND
Maximum Input Current	±1 mA
Output Current	20 mA
Junction Temperature	125°C
Continuous Pdiss (T = 85°C) (Derate 20.4 mW/°C above 85°C)	0.816 W
Thermal Resistance (Rth) (Junction to Lead)	49 °C/W
Storage Temperature	-65°C to +150°C
Operating Temperature	-40°C to +85°C

Bit Rate	5.00000 Gb/s
Pat Length	127 Bits
Div. Ratio	1:8

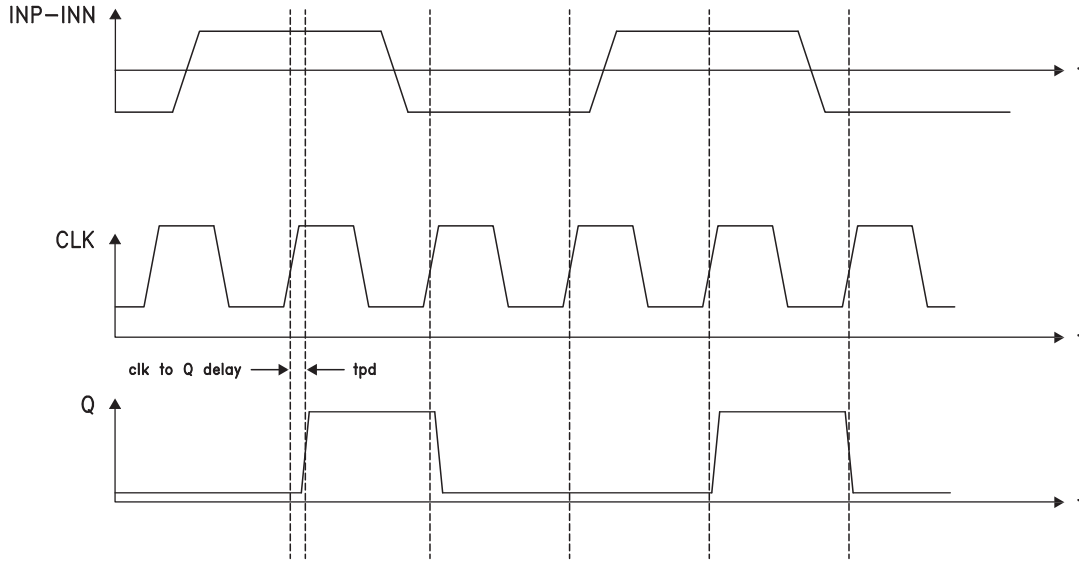


ELECTROSTATIC SENSITIVE DEVICE  
OBSERVE HANDLING PRECAUTIONS

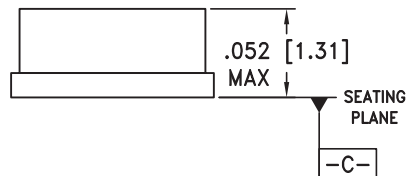
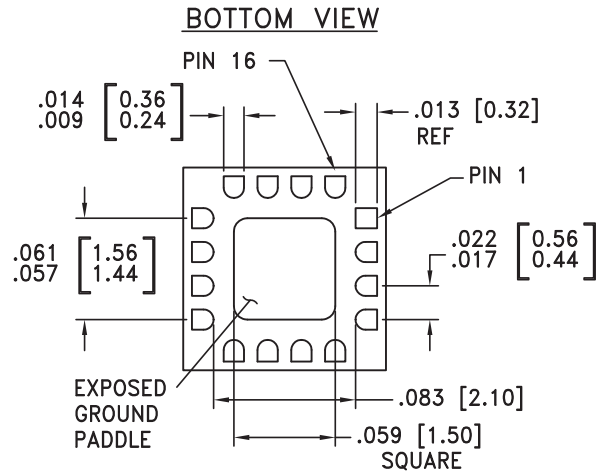
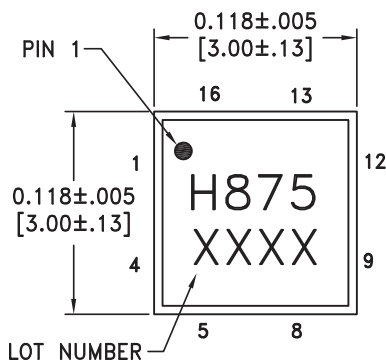


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### Timing Diagram



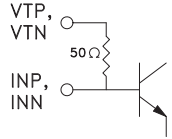
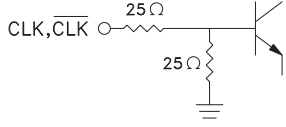
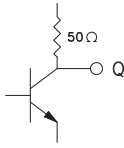
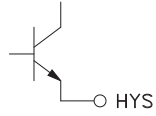
### Outline Drawing



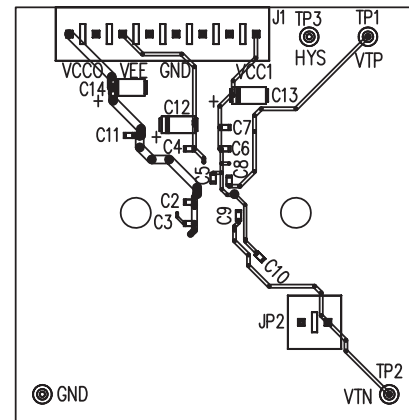
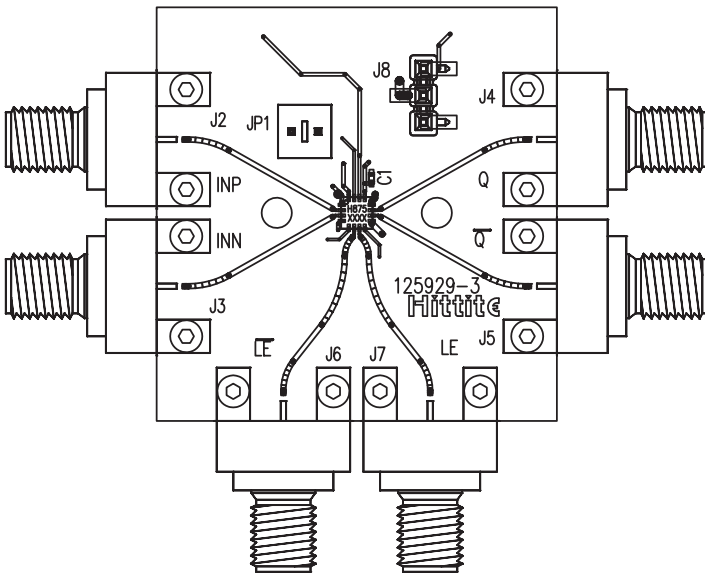
NOTES:

1. PACKAGE BODY MATERIAL: ALUMINA
2. LEAD AND GROUND PADDLE PLATING:  
30-80 MICROINCHES GOLD OVER 50 MICROINCHES MINIMUM NICKEL.
3. DIMENSIONS ARE IN INCHES (MILLIMETERS).
4. LEAD SPACING TOLERANCE IS NON-CUMULATIVE.
5. PACKAGE WARP SHALL NOT EXCEED 0.05mm DATUM -C-
6. ALL GROUND LEADS MUST BE SOLDERED TO PCB RF GROUND.
7. PADDLE MUST NOT BE DC GND. THERMAL DISSIPATION PATH ONLY.

**Pin Descriptions**

Pin Number	Function	Description	Interface Schematic
1	VTP	Termination resistor return pin for INP Input.	
2	INP	Non-Inverting analog input	
3	INN	Inverting analog input	
4	VTN	Termination resistor return pin for INN input	
5, 16	Vcci	Positive supply voltage input stage.	
6	$\overline{\text{CLK}}$	Clock input pin, inverting side.	
7	CLK	Clock input pin, non-inverting side.	
8	CLKRTN	Clock RTN pin, connect to GND.	
9, 12	Vcco	Positive supply voltage for the output stage.	
10	$\overline{\text{Q}}$	Inverting output. Q bar is at logic low if the analog voltage at the non-inverting input, INP, is greater than the analog voltage at the inverting input, INN, after a positive transition on CLK and negative transition on $\overline{\text{CLK}}$ .	
11	Q	Non-inverting output. Q is at logic high if the analog voltage at the non-inverting input, INP, is greater than the analog voltage at the inverting input, INN, after a positive transition on CLK and negative transition on $\overline{\text{CLK}}$ .	
14	HYS	Hysteresis Control pin. This pin should be left disconnected to minimize hysteresis. Connect to Vee with a resistor to add the desired amount of hysteresis.	
13	Vee	Negative power supply, -3V.	
15	RTN	Return for ESD protection, connect to GND.	
	Package Base	Do not DC GND. Thermal dissipation path only.	

### Evaluation PCB



### List of Materials for Evaluation PCB 125932 <sup>[1]</sup>

Item	Description
J1	8 Pos. Vertical TIN
J2 - J7	2.92 mm 40 GHz Jack
J8	Terminal Strip, Single Row 3 Pin SMT
JP1, JP2	2 Pos. Vertical TIN
C1 - C3, C5, C6, C8 - C10	100 pF Capacitor, 0402 Pkg.
C4, C7, C11	330 pF Capacitor, 0402 Pkg.
C11 - C13	4.7 uF Tantalum
TP1 - TP4	DC Pin, Swage Mount
U1	HMC875LC3C Comparator
PCB	125929 Evaluation PCB

[1] Reference this number when ordering complete evaluation PCB

[2] Circuit Board Material: Rogers 4350 or Arlon 25FR

The circuit board used in the final application should use RF circuit design techniques. Signal lines should have 50 ohm impedance while the package ground leads should be connected directly to the ground plane similar to that shown. The exposed paddle should not be electronically connected to DC GND, thermal dissipation path only. A sufficient number of via holes should be used to connect the top and bottom ground planes in order to provide good RF grounding to 25 GHz. The evaluation circuit board shown is available from Hittite upon request.

### Application Circuit

